Forth computer

Exceeding Brian Woodroffe's earlier expectations, his 6809-based Forth computer can be used with disc drives requiring high data-transfer rates — including Sony's microdrive and 8in floppy-disc drives. Access times for standard drives can also be reduced using a minor hardware modification.

After using the computer for a while I became discontented with the disc system because the software caused a one second delay each time access to the disc was required. This waiting time is needed to allow the drive to reach its operating speed - the software doesn't know whether the disc is running or stopped before access and much of the benefit of the virtual memory system is lost because of this delay. Forth keeps data from a number of disc sectors in memory. When data from a sector that is not in the main memory is requested by the program, Forth overwrites one of the buffers with the required data. But if data in the buffer has been changed Forth first sends the data back to the disc so there can be two 1s delays for one disc call.

Keeping the disc constantly rotating is the easiest way of avoiding this delay but this was rejected because it shortens the life of the drive even though the motor is a brushless d.c. type. The method chosen relies on the fact that disc access operations are not uniform in time i.e., they are likely to come in bursts, especially when loading or listing screens from a disc and as a result of the virtual-memory buffer replacement algorithm described above. Keeping the drive running for a short while after a disc access is made means that it is likely that the disc will be running when the next disc access is required.

Normally, the disc-drive motor is turned off by the disc-select signal going false (p.i.a. A port, D₆, 0 to 1) when the program finishes using the drive. In the modification the drive motor enable signal is held true for five seconds after the driveselect signal goes false by a monostable multivibrator triggered by the trailing edge of the p.i.a. signal. As the software always assumes that the drive is up to speed and available, even though the monostable i.c. might have completed its cycle, a means of ensuring that the WD1793 controller doesn't try to access the drive during the motor start period is required. During this period the ready signal is held false by a further monostable multivibrator fired by the drive-motor start signal. A low-pass filter after the NOR gate combines the two sources of motor-on signal to allow for the set-up time of the 5s monostable.

This small hardware modification, consisting of two s.s.i. devices relieves the software of all considerations of motor-start latency. To prevent erroneous trig-

Brian Woodroffe works in research and development at Hewlett Packard.

by B. Woodroffe

gering the two monostable multivibrators should be grounded separately.

Interfacing 8in drives

I found it galling that my 1.5MHz 6809 Forth system could not be interfaced with faster 8in drives, especially as these are often available second-hand at bargain prices. I have not yet got an 8in drive but I have been fortunate enough to try one of the sub-5in drives from Sony which has the same data rate as 8in drives. There is as yet no de jure standard for microfloppydisc drives but within Hewlett Packard, the Sony drive is the de facto standard. The first problem is to build a data-service routine that services the disc at a rate better than 11µs/byte. Although nominal discdata transfer normally takes 16µs/byte, Western Digital specify 11 and 13us worstcase service times for write and read respectively.

The previously used software loop (Wireless World, June 1983) achieves far worse than 11µs, even with the M6809 direct-page register modified to make the

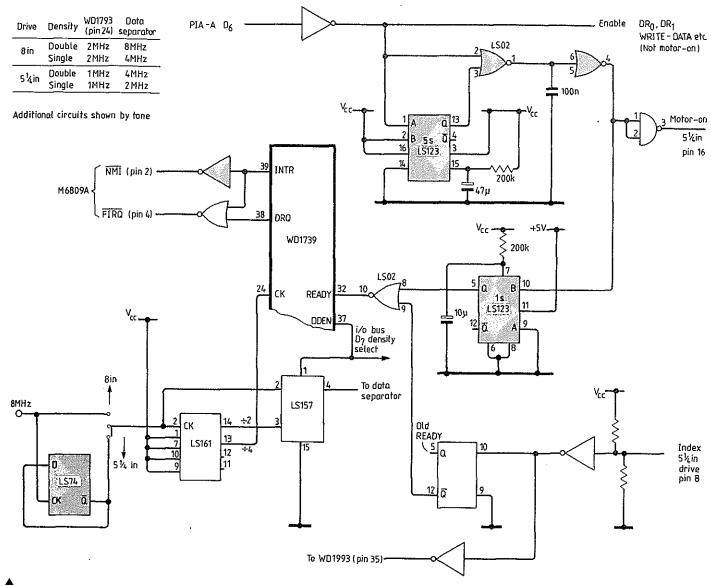
16809 to the total

List 1. In this design the following Forth words are available.

controller i.c. accessible through direct addressing. Analysing the software loop shows that two functions are being carried out - a byte is transferred between the WD1793 controller and ram, and bytes are counted to determine when the sector operation is completed. If the second function could be dispensed with the remaining loop would be much smaller and faster. There would be a small penalty in that the ability to read from and write to consecutive sectors would be lost as no byte count is kept. The problem now is how to break out of the disc-service software loop. Fortunately the controller gives hardware help here in that the IRQ line is activated on completion of every command; the DRQ line is activated for each byte transfer. DRO, connected to the M6809 FIRO line, is used in the data transfer loop to make the processor clear its SYNC state, thus synchronizing controller/processor transfers operations.

Interrupt request IRQ is tied to one of the other M6809 interrupt lines so that when a read or write-sector command is complete the processor aborts its current data-transfer loop activity and commences the interrupt routine. On application of the FIRQ signal the processor does not abort the data transfer loop and carry out the FIRQ routine because the program inhibits the FIRQ interrupt by holding the FIRQ mask bit in the condition-code re-

LIT	EXECUTE	BRANCH	DBRANCH	(LDDP)	(+L.002)
(00)	1	1	K	DIGIT	(FIND)
ENCLOSE	CHIT	KEY	7 TERMINAL	CR	CHOVE
uc	Uх	U/	AND	OR	XOR
SPE	SPI	RP i	; S	+	D+
ศากบร	DMINUS	1+	2+	1	2-
М×	×	87	/MOD	,	អាចជ
*/#DD	8/	M/NOD	ABS	DABS	S~)D
4	D+	Ð	0 <	LEAVE	≯R
R>	R	OVER	DROP	SWAP	ρÛΡ
+ 1	P	60		C:	TOGGLE
< BUILDS	DOES>	;	;	CONSTANT	VARIABLE
f)	1	3	ź	BL.	CVI.
FIRST	LUHIT	USER	*ORIGIN	50	R D
TIB	WIDTH	WARNING	FENCE	DP	AOC-1-INK
TIL K	Itt	OUT	SCR	OFFSET	CONTEXT
CURRENT	STATE	BASC	ĐPL.	FLD	CSP
R I	HL D	COLUMNS	HERE	ALLOT	ÇOF
,	ζ,	-		\ \	
>	SPACE	MIN	MAX	-DUP	
TRAVERSE	LATEST	LFA	CFA	NFA .	PFA
LOSP	FERROR	7C06P	7EXFC	7PA18S	7CSP
71. BADING	COMPILE	Ε	1	SMUDGE	HEX
DECIMAL	(GCODE)	COUNT	TYPE	-TRAILING	AEV.
("'	C. " *	. "	DEPTH	2STACK	?FREE
ROI	EXPECT	QUERY	X	FILL	ERASE
BLANKS	HOLD	PAD	์ พิติส ฮ	(NUMBER)	
FIND	(ABOR E)	ERROR	ID.	CREATE	NUMBER
LITERAL	DILITERAL	INTERPRET	INMEDIATE	VOCARULARY	(COMPILE)
	QUIT	ABORT	COLD	WARM	DEFINITIONS
USC	PREV	* BDF	(LINE)	.LINE	*DR
GAD)	. 000	FORGET		MESSAGE ·
ENDIF	THÉN	po	L00P	BACK	REGIN
IND.	AGAIN	REPEAT		+L80P	UNTIL
SPACES	(1	1)	TF STGN	ELSE	HHITE
. R	Ď.	17	SILN	1	D.R
90016	VI.IST	B/BUF		LIST	INDEX
080	081	ov nor	B/SCR	SCR > BLK	USEBLR
OPDATE	EMPTY-BUFFERS BLOCK		DRIVE	R/W	; CDDE
4005	WILLIAM LENS BEREK		FLUSH	FORTH	Task

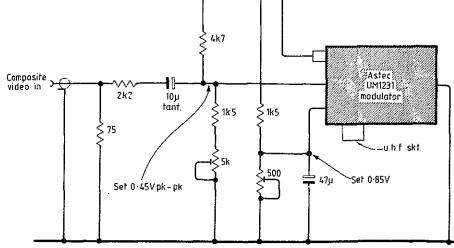


Disc interface modification speeds up overall access time by keeping the drive motor running for five seconds after the computer tells it to switch off. This significantly reduces the effect of a one second delay required for the motor to start up since disc-access operations tend to come in bursts.

U.h.f. modulator connects to the videocontroller circuit output (see June issue) so that the computer can be used with a standard tv set.

gister set. The controller interrupt-request line IRQ, being connected to the processor's non-maskable interrupt input NMI, can never be masked so when this line is true the processor must be interrupted and goes to the routine requested by IRQ. The processor IRQ interrupt-request input could have been used but I wanted to leave it free for expansion.

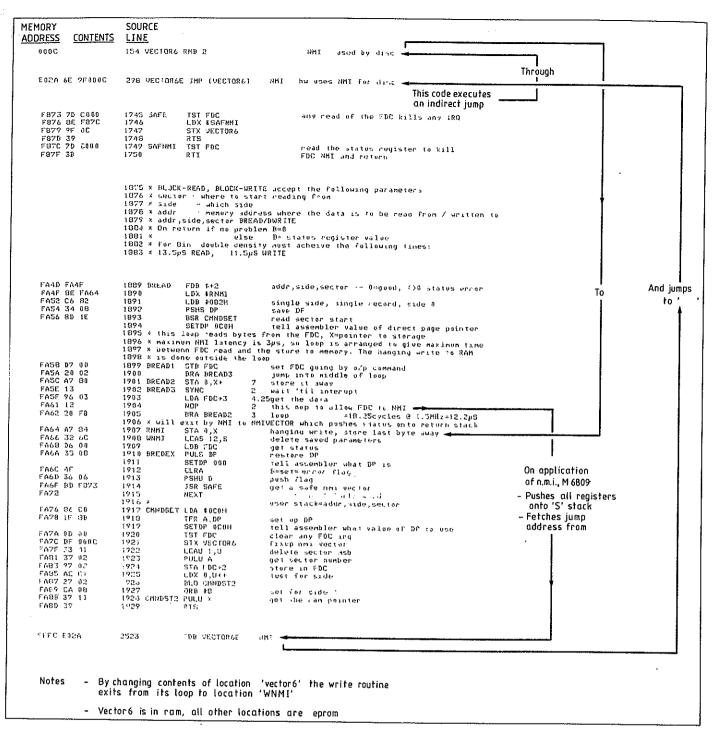
For sector read and write operations the disc controller interrupts on transfer of the last byte. In the case of a sector-write operation the data-transfer routine is finished when the last byte is written into the controller. In sector read operations the data transfer routine is finished not when the last byte is read from the controller, but when it is written into the ram sector buffer. So when a sector is written the



+51/

controller interrupts after all data transfers have taken place but when sectors are read the controller causes a jump out of the software loop before the last memory-storage operation is carried out. Worse still, latency before the controller interrupt is variable so when an interruption is made, whether or not the memory has been updated remains in doubt. The solution I chose was to code the data-transfer loop so as to maximize the time between reading data from the controller and writing it into

the memory. Inclusion of a no-operation, NOP, increases the data-transfer loop time to just under the allowable maximum of 13µs to ensure that the controller always interrupts before the processor can write the byte into memory; the first operation of the interrupt routine is to write that byte into memory. Unfortunately this writing operation done outside the data-transfer loop means that the interrupt routines for sector reading and writing must be different.



The processor starts its interrupt sequence by pushing appropriate registers onto the stack and jumping to code pointed to by a vector in high memory. In this Forth system high memory is eprom so the interrupt vectors cannot be changed to point to different routines. I remedied this by making the interrupt vectors point to code which executes a jump to a location determined by a value stored in ram. By changing data in the ram location the non-maskable interrupt vector can be altered during program execution. This practice is unstructured and therefore unfashionable, but it is highly effective.

Normally an interrupt routine is completed by a return-from-interrupt instruction which restores processor register values to those prior to the interrupt, i.e. restore context. In this case the interrupt vector is being used as a jump instruction to jump out of the data-transfer loop so the first operation in the NMI routine is to

Diagram of program flow during a sector read on the Forth computer.

delete saved registers (LEAS 12,8). As the controller is connected to the non-maskable interrupt line there is the potential for the occurrence of an interrupt when one is not required. To prevent this the NMI vector points to a safe routine when not in use which reads the controller status register, clearing the cause of the interrupt before carrying out a more normal return from interrupt, RTI, operation.

Extra signals to the disc drives, e.g. track 42, should be inverted and buffered using standard t.t.l. open-collector drivers (7438) as used for the WGATE signal. Extra input signals from the drive such as READY should be buffered using say two LS14 gates, and terminated as previously shown. I should have used the drive's ready signal, eliminating one of the monostable multivibrators connected to the index line but I did not. Also I connected my

motor-on signal (5.25in drive, pin16) to the Sony drive head-load line, pin 14, whereas I should have used hold, HLD on pin 28 of the controller. These minor modifications were made because I still intend to use 5.25in drives as they currently offer better value for money than the Sony drives at one-off prices.

The matter of write-precompensation has not yet been resolved. I found by trial and error that for the Sony drive at least write-precompensation is not mandatory. It might be necessary for older 8in drives, and in commercial products to minimize the number of attempts to read the disc. Details of precompensation circuits are given in the Western Digital handbook and reference two.

For drives that keep the disc rotating, such as Sony's and most 8in drives, the disc speed-up hardware previously described should not be fitted but the drive should be connected with the motor-on

List 2. Forth words specific to this system.

Note the movement of data onto and off the stack is shown thus s1 s2 ..ARC., r1 r2 r3

Forth word 'ABC' takes two values off the stack (s2=top of stack) and produces three results (r3=new top of stack).

VEHIT,VKEY,V7TER - a user variable containing the execution vector for EMIT,KEY, 7TERMINAL respectively ..VEHIT..addr ..VKEY..addr ..V?TER.,addr

DPMAX = a user variable containing the maximum address for the SHAX a user variable containing the maximum depth of data stack allowed for this user .SMAX..addr WORD to store a byte to, read a byte from one of the PI,PR a WORD to store a byte to, read a byte from one of the user ports data addr..Pt..
 a constant =0 for single density; -1 for double density .DENSITY..boolean
 a WORD to select the drive value top of stack value..DR-SEL..
 a WORD to deslect the disc drives ..DE-SEL..
 a variable which if true causes a read after write purity of discrepands. DENSITY DR-SEE DF-SE VERTEY SIDE/DISC SEC/TRK.TRK/SIDE SIDE PURITE CMND RATE SEEK stack of stack
value ..SEEK..
a WORD to step*the flappy disc, IN if TOS=1
OUT if TOS=-1
same direction if TOS=6 STEP value ..SEEK..
reads the address mark off the disc, either returns
a/ flag=true if unsuccessful (-status of WD1793) R-ADR b/ sector_type, sector, side, track , false flag .R-ADR.. type sector side track 0 or .R-ADR.. true sector side track 0 sector side track 10 or .R-ADR.. true similar to FORTH BLOCK_READ, BLOCK_WRITE except additional parameters are included, sector within a track PREAD BURITE track,
returns WD1793 status 0-successful
address side sector .HREAD..status
similar to FORTH R/W except no limit (hecking is done.
Returns to top of stack a flag. 0-successful operation
1-crt error
22-cct parts. SEC-R/W addr sector flag .SEC-R/W. Status a WDRD to write a whole track to disc, used in formatting a disc. Takes from the stack the address of the track change. Return WD1793 status, 0-secressful addr .TRK-WR. Status a WDRD to build in memory a byte image of a track prior to be written out by TRK-WP. Takes as input the side and track number whose image is to be formed and produces the address of where the image is. side track .TRK-WD addr a WDRD to find out what surt of disc is on the drive. sets CONSTANT R/BUI, DENSITY, SEC/TRK to suit. ... 70185... 3-not ready TRK WR TRK-BLI 201S0 sets CONSTANT's R/BUF, DENSITY, SEC/TRK to suit.
..7DISC..
a WORD to delay execution by the number of milliseconds that is top of stack value ..MS..
everently a no operation word, that canbe patched to allow interleaving of the sectors when formatting a disc. Intended usuage is that it will modify the sector number currently at top of stack.
sector in ..SEEW., sector in MS SICER

signal permanently true, i.e. grounded.

Software issued (first revision) assumes the presence of disc speed-up hardware and includes the faster data-transfer loop. I will supply a drive pin connection list and format program for the Sony drive that can be modified for 8in drives to readers sending an s.a.e. to me at 632 Queensferry Road, Edinburgh. The Forth word BLD-TRK in eprom is only suitable for minifloppy disc drives.

Thanks to Hewlett Packard for the use of their test equipment and Sony for the loan of a microdrive. Software used, based on the FIG model, was prepared on an HP64000 microprocessor development system.

Integrated circuits 87 and 88 were missing from last month's components list. They are 2114 static rams. In the photo-

graph of power-supply spikes, vertical sensitivity for all but the clock signal is 0.5V/div.

References

J. Bovin, Floppy incompatibility, Systems International, May 1983, p.61.
 J. Hoeppner and L. Wall, Encoding/decoding techniques double floppy disc capacity, Computer Design, Feb. 1980, pp. 127-135.

Brian Woodroffe plans to describe the Forth language in a subsequent series.

Wireless World Forth computer Introduction, c.p.u. and memory circuits, May 1983, pp. 53-8.
Circuit description, video-controller circuit and peripherals, June 1983, pp. 55-8.
Software, disc controller and power supply circuits, July 1983, pp. 58-61.
Construction tips, August 1983, pp.44,45.

Complementary current mirror

Current mirrors with transistors of the same type of conductance are well known¹ and widely used in integrated circuits². It is possible to create the configuration with similar properties using complementary transistors also, Fig. a. Accepting the usual assumptions² that

$$\begin{split} I_{C1} &= I_{S1} exp |V_{BEI}| / V_{T} \\ I_{C2} &= I_{S2} exp V_{BE2} / V_{T} \\ I_{B1} &= I_{C1} / \beta_{F1} \end{split}$$

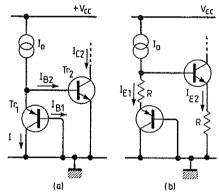
and $I_{B2}=I_{S2}/\beta_{F2}$, the output current is

$$I_{C2} = \frac{I_0}{\left(1 + \frac{1}{\beta_{F1}}\right) \frac{I_{S1}}{I_{S2}} + \frac{1}{\beta_{F2}}}$$

with $|V_{BEI}|=V_{BE2}$. If the technology allows two complementary transistors with $I_{SI}=I_{S2}$, then

$$I_{C2} = \frac{I_0}{1 + \frac{1}{\beta_{F1}} + \frac{1}{\beta_{F2}}} \approx I_0 \left(1 - \frac{1}{\beta_{F1}} - \frac{1}{\beta_{F2}} \right)$$

as in the ordinary current mirror. Usually n-p-n and p-n-p transistors in an integrated technology are produced by different methods and parameters I_{S1} and I_{S2}



Complementary transistor current mirror with matched transistor (a) and matched emitter resistances (b).

are not matched. But the discrete current mirror with matched resistors in emitter circuits works reasonably well, Fig. b. In this circuit

$$I_{E1}R + |V_{BE1}| = V_{BE2} + I_{E2}R$$

and

$$I_{E1} {=} I_{E2} {+} \frac{|V_{BE1}| {-} V_{BE2}}{R} {\cdot}$$

If transistors are designed for complementary operation, say 2N4401 and 2N4403, and emitter resistors are matched to within 1% the error is 2 to 3% without preliminary transistor matching. The gain β_{FI} of a discrete p-n-p transistor is usually high and the collector currents happen to be matched also. - I. M. Filanovsky, University of Alberta.

1. F.J. Lidgey. Looking into current mirrors. Wireless World, October, 1979, vol. 68, pp. 51-58.

2. P. Gray, R. Meyer. Analysis and design of analog integrated circuits. Wiley, 1977.